

Solutions to Chapter 10 and 11 Exercises

Problem 10.8

Let

$$s(t) = \sum_{k=-\infty}^{\infty} s_k \exp(j2\pi f_o t).$$

Then

$$s(t - T) = \sum_{k=-\infty}^{\infty} s_k \exp(j2\pi f_o(t - T))$$

$$R_{X,X}(\tau) = E \left[\sum_k \sum_m s_k s_m^* \exp(j2\pi k f_o(t - T)) \exp(-j2\pi m f_o(t + \tau - T)) \right]$$

$$= \sum_k \sum_m s_k s_m^* \exp(j2\pi(k - m)f_o t) \exp(-j2\pi m f_o \tau)$$

$$E[\exp(j2\pi(k - m)f_o T)]$$

$$E[\exp(j2\pi(k - m)f_o T)] = \frac{1}{t_o} \int_0^{t_o} \exp(j2\pi(k - m)f_o u) du$$

$$= \begin{cases} 0 & k \neq m \\ 1 & k = m \end{cases}$$

$$R_{X,X}(\tau) = \sum_{k=-\infty}^{\infty} |s_k|^2 \exp(-j2\pi k f_o \tau)$$

$$S_{X,X}(f) = \sum_{k=-\infty}^{\infty} |s_k|^2 \delta(f - k f_o)$$

Hence the process $X(t)=s(t-T)$ has a line spectrum and height of each line is given by the magnitude squared of the Fourier series coefficients.

Problem 10.12

$$R_{X,X}(t_1, t_2) = E[\cos(\omega_c t_1 + B[n_1]\pi/2) \cos(\omega_c t_2 + B[n_2]\pi/2)],$$

where n_1 and n_2 are integers such that $n_1 T \leq t_1 < (n_1 + 1)T$ and $n_2 T \leq t_2 < (n_2 + 1)T$. For t_1, t_2 such that $n_1 \neq n_2$,

$$R_{X,X}(t_1, t_2) = E[\cos(\omega_c t_1 + B[n_1]\pi/2)]E[\cos(\omega_c t_2 + B[n_2]\pi/2)] = 0,$$

while for t_1, t_2 such that $n_1 = n_2$,

$$\begin{aligned} R_{X,X}(t_1, t_2) &= \frac{1}{2} \cos(\omega_c(t_2 - t_1)) + \frac{1}{2} E[\cos(\omega_c(t_2 + t_1) + \pi B[n_1])] \\ &= \frac{1}{2} \cos(\omega_c(t_2 - t_1)) - \frac{1}{2} \cos(\omega_c(t_2 + t_1)) \end{aligned}$$

Since this autocorrelation depends on more than just $t_1 - t_2$, the process is not WSS.

(b) From part (a),

$$R_{X,X}(t, t+\tau) = \begin{cases} 0 & \text{if } t, t+\tau \text{ are in different intervals,} \\ \frac{1}{2} \cos(\omega_c \tau) - \frac{1}{2} \cos(\omega_c(2t + \tau)) & \text{if } t, t+\tau \text{ are in the same intervals.} \end{cases}$$

Since the process is not WSS we must take time averages.

$$R_{X,X}(\tau) = \langle R_{X,X}(t, t+\tau) \rangle = (1-p(\tau))\langle 0 \rangle + p(\tau) \langle \frac{1}{2} \cos(\omega_c \tau) - \frac{1}{2} \cos(\omega_c(2t + \tau)) \rangle,$$

where $p(\tau)$ is the fraction of the values of t that lead to t and $t + \tau$ being in the same interval. This function is given by

$$p(\tau) = \begin{cases} 0 & |\tau| > T, \\ 1 - \frac{|\tau|}{T} & |\tau| < T. \end{cases}$$

Therefore,

$$\begin{aligned} p(\tau) &= \text{tri}(t/T) \\ R_{X,X}(\tau) &= \frac{1}{2} \text{tri}(\tau/T) \cos(\omega_c \tau) \\ S_{X,X}(f) &= \frac{1}{2} FT[\text{tri}(\tau/T)] * FT[\cos(\omega_c \tau)] \end{aligned}$$

using Table E.1 in Appendix E in the text,

$$\begin{aligned} S_{X,X}(f) &= \frac{1}{4} T \text{sinc}^2(fT) * (\delta(f - f_c) + \delta(f + f_c)) \\ &= \frac{T}{4} (\text{sinc}^2((f - f_c)T) + \text{sinc}^2((f + f_c)T)) \end{aligned}$$

Problem 10.14

- (a) The absolute BW is ∞ since $S(f) > 0$ for all $|f| < \infty$.
 (b) The 3dB BW, f_3 satisfies

$$\frac{1}{(1 + (f_3/B)^2)^3} = \frac{1}{2}$$

$$\Rightarrow f_3 = B\sqrt{2^{1/3} - 1} = 0.5098B.$$

(c)

$$\int_{-\infty}^{\infty} f^2 S(f) df = \int_{-\infty}^{\infty} \frac{f^2}{(1 + (f/B)^2)^3} df = B^3 \int_{-\infty}^{\infty} \frac{z^2}{(1 + z^2)^3} dz = \frac{\pi}{8} B^3$$

$$\int_{-\infty}^{\infty} S(f) df = \int_{-\infty}^{\infty} \frac{1}{(1 + (f/B)^2)^3} df = B \int_{-\infty}^{\infty} \frac{1}{(1 + z^2)^3} dz = \frac{3\pi}{8} B$$

$$B_{rms}^2 = \frac{\frac{\pi}{8} B^3}{\frac{3\pi}{8} B} = \frac{B^2}{3}$$

$$B_{rms} = \frac{B}{\sqrt{3}}.$$

Problem 10.19

(a)

$$\begin{aligned} E[\epsilon^2] &= E[(Y[n+1] - a_1Y[n] - a_2Y[n-1])^2] \\ &= R_{Y,Y}[0](1 + a_1^2 + a_2^2) - 2a_1(1 - a_2)R_{Y,Y}[1] - 2a_2R_{Y,Y}[2] \end{aligned}$$

(b)

$$\begin{aligned} \frac{\partial E[\epsilon^2]}{\partial a_1} &= 2a_1R_{Y,Y}[0] - 2(1 - a_2)R_{Y,Y}[1] = 0 \\ &\Rightarrow R_{Y,Y}[0]a_1 + R_{Y,Y}[1]a_2 = R_{Y,Y}[1] \\ \frac{\partial E[\epsilon^2]}{\partial a_2} &= 2a_2R_{Y,Y}[0] - 2R_{Y,Y}[2] + 2a_1R_{Y,Y}[1] = 0 \\ &\Rightarrow R_{Y,Y}[1]a_1 + R_{Y,Y}[0]a_2 = R_{Y,Y}[2] \\ &\Rightarrow \begin{bmatrix} R_{Y,Y}[0]R_{Y,Y}[1] & R_{Y,Y}[0] \\ R_{Y,Y}[1] & R_{Y,Y}[0] \end{bmatrix} \begin{bmatrix} a_1 \\ a_2 \end{bmatrix} = \begin{bmatrix} R_{Y,Y}[1] \\ R_{Y,Y}[2] \end{bmatrix} \\ \Rightarrow \begin{bmatrix} a_1 \\ a_2 \end{bmatrix} &= \frac{1}{R_{Y,Y}[0] - R_{Y,Y}[1]} \begin{bmatrix} R_{Y,Y}[0]R_{Y,Y}[1] - R_{Y,Y}[1]R_{Y,Y}[2] \\ R_{Y,Y}[0]R_{Y,Y}[2] - R_{Y,Y}[1]^2 \end{bmatrix} \end{aligned}$$

Problem 11.10

$$\begin{aligned}
Y(t) &= a(S^2(t) + N^2(t) + 2S(t)N(t)) \\
E[Y(t)] &= aE[S^2(t)] + aE[N^2(t)] + 2aE[S(t)N(t)] \\
&= a(\sigma_S^2 + \sigma_N^2). \\
R_{Y,Y}(\tau) &= a^2E[(S^2(t) + N^2(t) + 2S(t)N(t)) \\
&\quad (S^2(t+\tau) + N^2(t+\tau) + 2S(t+\tau)N(t+\tau))] \\
&= a^2E[S^2(t)S^2(t+\tau)] + a^2E[N^2(t)N^2(t+\tau)] \\
&\quad + 4a^2E[S(t)S(t+\tau)N(t)N(t+\tau)] + a^2E[S^2(t)N^2(t+\tau)] \\
&\quad + 2a^2E[S^2(t)S(t+\tau)N(t+\tau)] + a^2E[N^2(t)S^2(t+\tau)] \\
&\quad + 2a^2E[N^2(t)N(t+\tau)S(t+\tau)] + 2a^2E[S(t)S^2(t+\tau)N(t)] \\
&\quad + 2a^2E[S(t)N(t)N^2(t+\tau)] \\
&= a^2[R_{S,S}^2(0) + 2R_{S,S}^2(\tau) + R_{N,N}^2(0) + 2R_{N,N}^2(\tau) \\
&\quad + 4R_{S,S}(\tau)R_{N,N}(\tau) + R_{S,S}(0)R_{N,N}(0) + R_{N,N}(0)R_{S,S}(0)] \\
&= a^2[(R_{S,S}(0) + R_{N,N}(0))^2 + 2(R_{S,S}(\tau) + R_{N,N}(\tau))^2] \\
&= a^2[(\sigma_S^2 + \sigma_N^2)^2 + 2(R_{S,S}(\tau) + R_{N,N}(\tau))^2].
\end{aligned}$$

In above Calculation we used the following information:

For two Gaussian random process X(t) and Y(t) :

$$\begin{aligned}
E[X^2(t)Y^2(t+\tau)] &= R_{XX}[0]R_{YY}[0] + 2R_{XY}^2[\tau] && \text{If X and Y are correlated} \\
E[X(t)Y(t)Y(t+\tau)] &= E[X(t)]E[Y(t)Y(t+\tau)] && \text{If X and Y are not correlated}
\end{aligned}$$

Problem 11.11

For the given PSD's,

$$\begin{aligned}
 R_{S,S}(\tau) &= F^{-1}[S_{S,S}(f)] = \frac{A^2}{4}(e^{-j2\pi f_c\tau} + e^{j2\pi f_c\tau}) = \frac{A^2}{2} \cos(2\pi f_c\tau) \\
 R_{N,N}(\tau) &= F^{-1}[S_{N,N}(f)] = \frac{N_o B}{2} \text{sinc}(B\tau)(e^{-j2\pi f_c\tau} + e^{j2\pi f_c\tau}) \\
 &= N_o B \text{sinc}(B\tau) \cos(2\pi f_c\tau)
 \end{aligned}$$

From these we determine that

$$\begin{aligned}
 \sigma_S^2 &= R_{S,S}(0) = \frac{A^2}{2}, \\
 \sigma_N^2 &= R_{N,N}(0) = N_o B.
 \end{aligned}$$

The autocorrelation of $Y(t)$ is then

$$\begin{aligned}
 R_{Y,Y}(\tau) &= a^2[(\sigma_S^2 + \sigma_N^2)^2 + 2(R_{S,S}(\tau) + R_{N,N}(\tau))^2] \\
 &= a^2 \left[\left(\frac{A^2}{2} + N_o B \right)^2 + 2 \left(\frac{A^2}{2} + N_o B \text{sinc}(B\tau) \right)^2 \cos^2(2\pi f_c\tau) \right] \\
 &= a^2 \left[\left(\frac{A^2}{2} + N_o B \right)^2 \right. \\
 &\quad \left. + 2 \left(\frac{A^2}{4} + A^2 N_o B \text{sinc}(B\tau) + (N_o B)^2 \text{sinc}^2(B\tau) \right) (1 + \cos(4\pi f_c\tau)) \right]
 \end{aligned}$$

Taking FT's, the PSD of $Y(t)$ is then

$$\begin{aligned}
 S_{Y,Y}(f) &= a^2 \left[\left(\frac{A^2}{2} + N_o B \right)^2 \delta(f) + 2 \left(\frac{A^2}{4} \delta(f) + A^2 N_o \text{rect}(f/B) + N_o^2 B \text{tri}(f/B) \right) \right. \\
 &\quad \left. * \left(\delta(f) + \frac{1}{2} \delta(f - 2f_c) + \frac{1}{2} \delta(f + 2f_c) \right) \right] \\
 &= a^2 \left(\frac{A^2}{2} + A^2 N_o B + (N_o B)^2 \right) \delta(f) + A^2 N_o \text{rect}(f/B) + N_o^2 B \text{tri}(f/B) \\
 &\quad + \frac{A^2}{4} (\delta(f - 2f_c) + \delta(f + 2f_c)) + A^2 N_o \left(\text{rect} \left(\frac{f - 2f_c}{B} \right) + \text{rect} \left(\frac{f + 2f_c}{B} \right) \right) \\
 &\quad + N_o^2 B \left(\text{tri} \left(\frac{f - 2f_c}{B} \right) + \text{tri} \left(\frac{f + 2f_c}{B} \right) \right).
 \end{aligned}$$

Problem 11.14

$$S_{Y,Y}(f) = S_{X,X}(f)|H(f)|^2 = \frac{a}{1 + (f/f_o)^2} \cdot b^2 \left[\text{rect} \left(\frac{f - f_c}{f_2 - f_1} \right) + \text{rect} \left(\frac{f + f_c}{f_2 - f_1} \right) \right],$$

where $f_c = (f_1 + f_2)/2$. Assuming $f_2 - f_1 \ll f_o$, the input PSD will be approximately constant over the passband of the filter. In that case, the output PSD will be well approximated by

$$\begin{aligned} S_{Y,Y}(f) &\approx S_{X,X}(f_c)|H(f)|^2 = \frac{ab^2}{1 + (f_c/f_o)^2} \cdot \left[\text{rect} \left(\frac{f - f_c}{f_2 - f_1} \right) + \text{rect} \left(\frac{f + f_c}{f_2 - f_1} \right) \right] \\ \Rightarrow &= \frac{2ab^2(f_2 - f_1)}{1 + (f_c/f_o)^2} \text{sinc}((f_2 - f_1)\tau) \cos(\omega_c\tau). \end{aligned}$$

Problem 11.26

$h(t)=s(t_0-t)$. In this case, $s(t_0-t)=s(t)$ so the impulse response of the matched filter is the same as the signal itself.