# L14: Optimal linear filtering - Wiener filtering

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## Reviewing L11-L13

What have we done so far?

- Signal models (L11-L12)
  - Nonparametric models: ACF and PSD.
  - Parametric models: AR, MA and ARMA.
- Signal model estimation (L13)
  - Nonparametric spectral estimation: the periodogram. **Pros:** 
    - fast to compute
    - asymptotically unbiased.

Cons:

- limited resolution for finite *N*:
  - $\leadsto$  the modified periodogram improves this
- large variance for all N:
  - $\rightsquigarrow$  Blackman-Tukey's method lowers variance.
- Parametric spectral estimation: AR-estimation.
  - **1** Estimate  $r_x[k]$  from data.
  - 2 Reformulate Yule-Walker to get  $\hat{a} = R_x^{-1} r_x$ .

## Learning objectives

After today's lecture you should be able to

- explain what type of problems Wiener-filters can solve.
- derive the Wiener-Hopf (WH) equations.
- use the WH-equations to derive a causal FIR Wiener filter.
- use the WH-equations to derive a **non-causal IIR Wiener** filter.
- Compute the mean squared error (MSE) of a Wiener-filter.

• Let *s*[*n*] and *w*[*n*] be zero mean, wide sense stationary processes and

$$x[n] = s[n] + w[n].$$

#### Objective

• Select H(z) to make e[n] as "small" as possible

$$x[n] \longrightarrow H(z) \xrightarrow{\hat{d}[n]} e[n] = \hat{d}[n] - d[n]$$
$$d[n]$$

• Small could mean different things. We use mean squared error

$$E\left\{e[n]^2\right\},$$

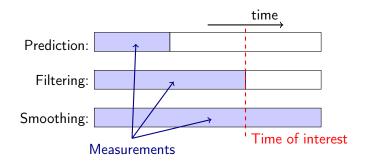
since this is easy to minimize.

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Based on measurements collected up until now, we encounter three common problems (k > 0):

- Filtering estimating <u>current</u> signal values, d[n] = s[n]. *Applications:* positioning, control systems, noise or echo cancellation, etc.
- Smoothing estimating past signal values, d[n] = s[n k]. Applications: signal analysis, image processing, system identification (modelling).
- **Prediction** estimating <u>future</u> signal values, d[n] = s[n + k]. *Applications:* decision making, planning, weather forecasts, etc.

These problems can be illustrated as



• We seek a linear estimator (filter)

$$\hat{d}[n] = h[n] \star x[n] = \sum_{k} h[k]x[n-k]$$

of *d*[*n*].

 As mentioned above, we wish to minimize the mean square error (MSE),

$$\mathsf{MSE}(\mathbf{h}) = E\left\{ \left( d[n] - \sum_{k} h[k] x[n-k] \right)^2 \right\}$$

where the vector **h** contains the impulse response coefficients h[k].

• The resulting Wiener filter  $\hat{d}[n]$  is a linear minimum mean square error (LMMSE) estimator.

## Wiener-Hopf (W-H) equations

- The W-H equations are *very important* and can be used to solve all the problems mentioned above.
- Objective: (again) We wish to minimize

$$\mathsf{MSE}(\mathbf{h}) = E\left\{\left(d[n] - \sum_{k} h[k]x[n-k]\right)^{2}\right\}$$

with respect to h.

- Derivation 1: the function is quadratic in h  $\Rightarrow$  it is convex in h
  - $\Rightarrow$  no local optima (except for the global optimum)
  - $\Rightarrow$  sufficient to differentiate and set to zero!

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#### Wiener-Hopf equations

## Wiener-Hopf (W-H) equations

• Differentiate the MSE:

$$\frac{\partial}{\partial h[t]} \mathsf{MSE}(\mathbf{h}) = \frac{\partial}{\partial h[t]} E\left\{ \left( d[n] - \sum_{k} h[k]x[n-k] \right)^{2} \right\}$$
$$= E\left\{ 2\left( d[n] - \sum_{k} h[k]x[n-k] \right) \left( -x[n-t] \right) \right\}$$
$$= -2r_{dx}[t] + 2\sum_{k} h[k]r_{x}[t-k]$$

• Setting this derivative to zero gives the

Wiener-Hopf (WH) equations  $\sum_{k} h[k]r_{x}[t - k] = r_{dx}[t],$ for all t where h[t] is free to select. Problem formulation General solution Filtering solutions

Causal FIR filters MSE of optimal FIR filter

## **FIR** filters

• Suppose H(z) is a causal FIR filter:

$$\hat{d}[n] = \sum_{n=0}^{p-1} h[k] x[n-k].$$

• The W-H eq's can be written as

$$\underbrace{\begin{bmatrix} r_{x}[0] & r_{x}[1] & \dots & r_{x}[p-1] \\ r_{x}[1] & r_{x}[0] & \dots & r_{x}[p-2] \\ \vdots & \vdots & \ddots & \vdots \\ r_{x}[p-1] & r_{x}|p-2] & \dots & r_{x}[0] \end{bmatrix}}_{\mathbf{R}_{x}} \underbrace{\begin{bmatrix} h[0] \\ h[1] \\ \vdots \\ h[p-1] \end{bmatrix}}_{\mathbf{h}} = \underbrace{\begin{bmatrix} r_{dx}[0] \\ r_{dx}[1] \\ \vdots \\ r_{dx}[p-1] \end{bmatrix}}_{\mathbf{r}_{dx}}$$

which yields that

$$\mathbf{h}_{\mathsf{opt}} = \mathbf{R}_x^{-1} \mathbf{r}_{dx}.$$

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## What is the minimum MSE?

 $\bullet$  The minimum MSE can be calculated by plugging in  $h_{\rm opt}:$ 

$$E\left\{e_{\min}^{2}[n]\right\} = E\left\{e_{\min}[n]\left(d[n] - \hat{d}_{opt}[n]\right)\right\} = \left\{Note: \ \hat{d}_{opt} \perp e[n]\right\}$$
$$= E\left\{\left(d[n] - \sum_{k=0}^{p-1} h_{opt}[k]x[n-k]\right)d[n]\right\}$$
$$= r_{d}[0] - \sum_{k=0}^{p-1} h_{opt}[k]r_{dx}[k] = r_{d}[0] - \mathbf{r}_{dx}^{T}\mathbf{R}_{x}^{-1}\mathbf{r}_{dx}$$

- Special case: if d[n] and x[n] are uncorrelated, then  $\hat{d}[n] = 0$  and the MSE is  $r_d[0]$ .
- In general, the more correlated (similar) x[n] is to d[n] the better is the estimate d
  <sup>^</sup>[n]!

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